

1.5MHz, 2A, VFB=0.6V Synchronous

Step-Down Converter

❖ GENERAL DESCRIPTION

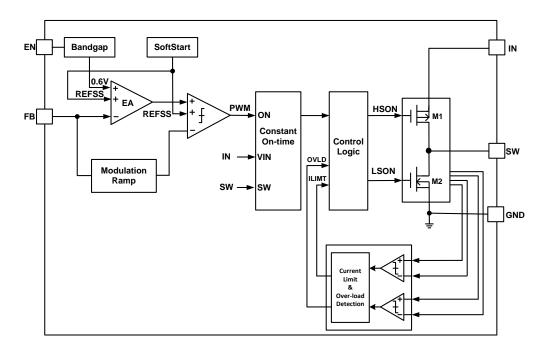
AX3702 is a high efficiency monolithic synchronous buck regulator using a constant on-time (COT) architecture. The device is available in an adjustable version. Supply current with no load is about 30uA and drops to <1uA in shutdown. The 2.5V to 6.0V input voltage range makes AX3702 ideally suited for single Li-lon, two to four AA battery-powered applications. 100% duty cycle provides low dropout operation, extending battery life in portable systems. The constant on-time control scheme simplifies loop compensation and offers excellent load transient response. The high gain error amplifier in the control loop provides excellent load and line regulation. Proprietary adaptive on-time helps AX3702 to achieve nearly constant switching frequency across load range. AX3702 has cycle-by-cycle current limit and hiccup mode to protect over-load or short circuit fault conditions. Switching frequency is internally set at 1.5MHz, allowing the use of small surface mount inductors and capacitors. The internal synchronous switch increases efficiency and decreases need of an external Schottky diode. Low output voltages are easily supported with the 0.6V feedback reference voltage. AX3702 is available in small SOT-23-5L, SOT-23-6L, TDFN-10L (3*3), and SOP-8L-EP packages.

❖ FEATURES

- 2.5V to 6.0V Input Voltage Range
- Output Voltage from 0.6V to V_{IN}
- High Efficiency: Up to 92%
- 1.5MHz Switching Frequency Operation
- 2A Output Current
- Quiescent Current: 30uA (input < 4.2V)
- No Schottky Diode Required
- 100% Duty Cycle in Dropout
- 0.6V Reference Allows Low Output Voltages
- Internal 1msec Soft-Start
- Proprietary Fast Transient Constant On Time Architecture Stable with low ESR Ceramic **Output Capacitors**
- Cycle-by cycle Current limit Protection
- Hiccup Mode for Short Circuit and Over-Load Protect
- ≤ 1µA Shutdown Current
- SOT-23-5L, SOT-23-6L, TDFN-10L (3*3), and SOP-8L-EP Pb-Free packages
- RoHS and Halogen free compliance.

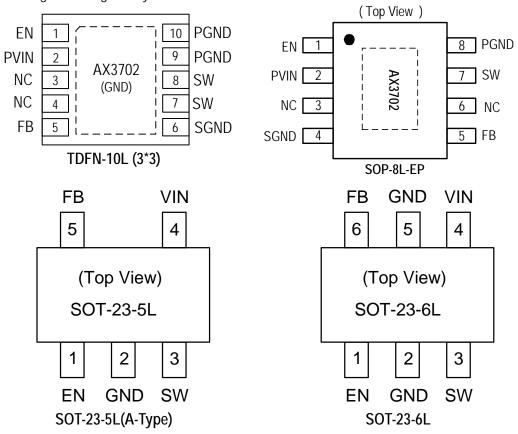


❖ BLOCK DIAGRAM



❖ PIN ASSIGNMENT

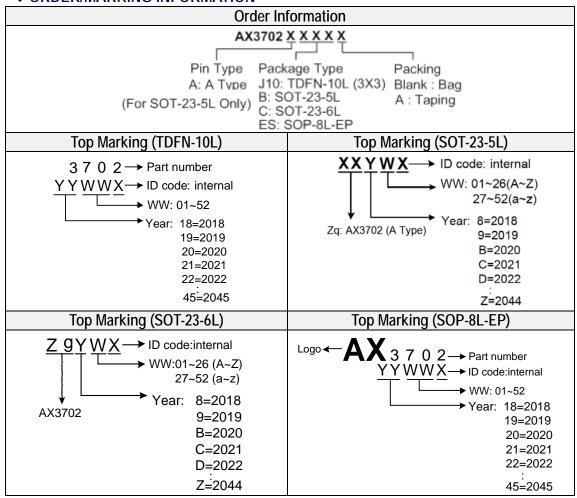
The packages of AX3702 are SOT-23-5L, SOT-23-6L, TDFN-10L (3*3), and SOP-8L-EP the pin assignment is given by:





EN	Enable pin H: normal operation L: Shutdown
PVIN	Power Supply Input Pin
SVIN	Signal Supply Input Pin
SGND	Signal Ground Pin
FB	Output Feedback pin
SW	Switch output pin. Connect external inductor here. Minimize trace area at this pin to reduce EMI.
PGND	Power Ground Pin
GND	Ground Pin
VIN	Power Supply Input Pin
NC	No Connect

❖ ORDER/MARKING INFORMATION



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❖ ABSOLUTE MAXIMUM RATINGS (at T_A = 25°C)

Characteristics		Symbol	Rating	Unit
VIN Pin Voltage		V_{IN}	V_{SS} - 0.3 to V_{SS} + 7	V
2ms Pulse width VIN tolerance		V _{PLUSE}	V_{SS} - 0.3 to V_{SS} + 10	V
Feedback Pin Voltage		V_{FB}	V_{SS} - 0.3 to V_{CC} + 0.3	V
EN Pin Voltage		V_{EN}	V_{SS} - 0.3 to V_{CC} + 0.3	V
Switch Pin Voltage		V_{SW}	V_{SS} - 0.3 to V_{CC} + 0.3	V
Peak SW Sink and Source Current		I_{PSW}	2.4	Α
Power Dissipation		PD	(T _J -T _A) / θ _{JA}	mW
Storage Temperature Range		T_{ST}	-65 to +150	°C
Operating Temperature Range		T _{OP}	-40 to +85	°C
Junction Temperature		ΤJ	+125	°C
ESD HBM		HBM	±2	KV
ESD MM		MM	±200	V
	TDFN-10L(3x3)		7.8	
Thermal Resistance from Junction	SOT-23-6L	θις	130	°C/W
to case	SOT-23-5L	O JC	130	C/VV
	SOP-8L-EP		15	
	TDFN-10L(3x3)		105	
Thermal Resistance from Junction	SOT-23-6L	Δ	250	°C/W
to ambient	SOT-23-5L	θ_{JA}	250	C/VV
	SOP-8L-EP		75	

Note: θ_{JA} is measured with the PCB copper area of approximately 1 in²(Multi-layer). That need connect to GND pin of the AX3702.



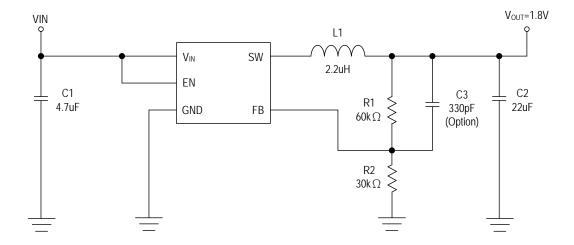
❖ ELECTRICAL CHARACTERISTICS

 $(V_{IN} = V_{EN} = 3.6V, T_A = 25$ °C, unless otherwise specified) (Note)

Characteristics	Symbol	Conditions	Min	Тур	Max	Units
Input Voltage Range	V _{IN}		2.5	-	6.0	V
Under Voltage Lock Out	V_{UVLO}	V _{IN} Rising	2.3	2.4	2.5	V
threshold	V UVLO	Hysteresis		280		mV
Input Over Voltage Lockout Threshold	$V_{ m OVLO}$	V _{IN} Rising		6.2		V
Input Over Voltage Lockout Hysteresis	OVEO	Hysteresis		400		mV
Feedback Voltage	V_{FB}	$V_{FB=}0.6V$	0.5880	0.6000	0.6120	V
Feedback Bias Current	I_{FB}	V _{FB} =0.65V	-30	-	+30	nA
Quiescent Current	Icc	V _{FB} =1V	-	30	-	uA
Shutdown Supply Current	I _{SD}	V _{EN} =0V	-	0.1	1	uA
Switching Current Limit	I _{LIMIT}	I _{OUT} =2A	2.4	2.8	3.2	Α
Line Regulation	$\Delta V_{\text{OUT}}/V_{\text{OUT}}$	$V_{IN} = 2.5V \sim 5.5V$	-	0.04	0.4	%/V
Load Regulation	$\Delta V_{OUT}/V_{OUT}$	I _{OUT} = 0.01 to 2A	-	0.5	0.6	%
Oscillation Frequency	Fosc	SW pin	1.2	1.5	1.8	MHz
R _{DS(ON)} of P-CH MOSFET	R _{DSON}	I _{OUT} =2A	-	0.19	-	Ω
R _{DS(ON)} of N-CH MOSFET	R _{DSON}	I _{OUT} =2A	-	0.10	-	Ω
Efficiency	EFFI	V _{IN} =5V, V _{OUT} =3.3V, I _{OUT} =0.5A	92	-	-	%
EN pin logic input threshold	V_{ENL}		-	-	0.4	V
voltage	V_{ENH}		1.0	-	-	V
EN Internal Pull Down Resistor	R_{EN}		700	1000	1300	kΩ
Soft-Start Interval	T _{SS}			1.0		ms
Thermal Shutdown Temperature	T _{SD}		-	160	-	°C
Thermal Shutdown Hysteresis	T_{SH}		-	30	-	°C
Output Discharging Resistor	R _{DISC}			50		Ω

Note: 100% production test at +25°C. Specifications over the temperature range are guaranteed by design and characterization.

❖ APPLICATION CIRCUIT



❖ FUNCTION DESCRIPTIONS

Theory of Operation

AX3702 is a constant on-time control synchronous step-down converter that offers excellent transient response over a wide range of input voltage. It achieves superior light-load efficiency with extremely low quiescent current.

Constant On-time Control

Constant on-time control step-down converters turn on HS immediately when FB droops below reference. The HS is turned on for a pre-determined period (on-time) of time to ramp up the inductor current, and then the LS will be turned on to ramp down the inductor current. The cycle repeats itself if FB droops below reference again. AX3702 uses proprietary technique to take into account the load current impact and adjusts the on-time accordingly to achieve a constant switching frequency over entire load current range.

For AX3702, the on-time is approximately:

$$T_{ON} = \frac{V_{OUT}}{V_{IN}} \cdot 0.66\mu$$

Due to its immediate response on FB voltage droop and simplified loop compensation, constant on-time offers a superior transient response compare to traditional fixed frequency PWM control step-down converters.

Light Load Operation

In light load condition where the converter operates in discontinuous mode, AX3702 cuts down its quiescent current to as low as 15uA and achieves excellent light load efficiency.

Enable

When input voltage is above the under voltage lock-out threshold, AX3702 can be enabled by pulling the EN pin to above 1V. AX3702 is disabled if the EN pin is pulled below 0.4 V.

Soft Start

AX3702 has built-in soft start of 1ms. During the soft start period, output voltage is ramped up linearly to the regulation voltage, independent of the load current level and output capacitor value.

Current Limit and Hiccup Mode

AX3702 has cycle-by-cycle HS current limit protection to prevent inductor current from running away. Once HS current limit is triggered, AX3702 will turn on LS and wait for the inductor to drop down to a pre-determined level before the HS can be turned on again. If this current limit condition is repeated for a sustained long period of time, AX3702 will consider it over-load or short circuit. Either way, AX3702 will enter hiccup mode, where it stop switching for a pre-determined period of time before automatically re-try to start up again. It always starts up with soft-start to limit inrush current and avoid output overshoot.

❖ APPLICATION INFORMATION

Setting the Output Voltage

Application circuit item shows the basic application circuit with AX3702 adjustable output version. The external resistor sets the output voltage according to the following equation:

$$V_{OUT} = 0.6V \times (1 + \frac{R1}{R2})$$

Table 1 Resistor select for output voltage setting

V _{OUT}	R2	R1
1.2V	30K	30K
1.5V	30K	45K
1.8V	30K	60K
2.5V	30K	95K
3.3V	30K	135K

Inductor Selection

For most designs, the AX3702 operates with inductors of 1μ H to 4.7μ H. Low inductance values are physically smaller but require faster switching, which results in some efficiency loss. The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_{L} \times f_{OSC}}$$

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Where is inductor Ripple Current. Large value inductors lower ripple current and small value inductors result in high ripple currents. Choose inductor ripple current approximately 35% of the maximum load current.

Table 2 Inductor select for output voltage setting ($V_{CC}=5V$)

V_{OUT}	1.2V	1.5V	1.8V	2.5V	3.3V
Inductor	2.2uH	2.2uH	2.2uH	2.2uH	2.2uH

Note: Part type MH or M (www.we-online.com)

For output voltages above 2.0V, when light-load efficiency is important, the minimum recommended inductor is 2.2 μ H. For optimum voltage-positioning load transients, choose an inductor with DC series resistance in the $50m\Omega$ to $150m\Omega$ range. For higher efficiency at heavy loads (above 200mA), or minimal load regulation (but some transient overshoot), the resistance should be kept below $100m\Omega$. The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation.

Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency shall be less than input source impedance to prevent high frequency switching current passing to the input. A low ESR input capacitor sized for maximum RMS current must be used. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. A $4.7\mu F$ ceramic capacitor for most applications is sufficient.

Output Capacitor Selection

The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current.

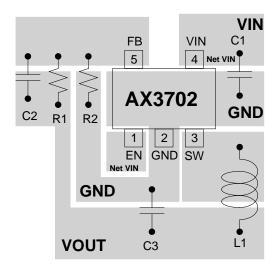
Compensation Capacitor Selection

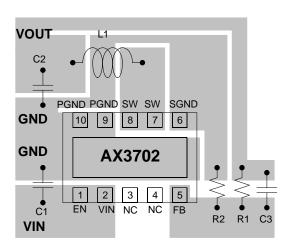
The compensation capacitors for increasing phase margin provide additional stability. It is required and more than 22pF, Refer to Demo Board Schematic, The optimum values for C3 (option) is 330pF.

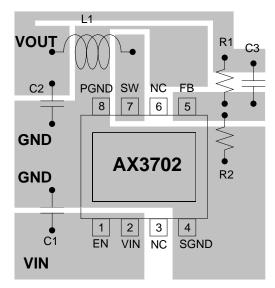
PCB Layout Recommendations

When laying out the printed circuit board, the following checking should be used to ensure proper operation of the AX3702. Check the following in your layout:

- 1. The power traces, consisting of the GND trace, the SW trace and the V_{IN} trace should be kept short, direct and wide.
- 2. Does the (+) plates of CIN connect to V_{IN} as closely as possible. This capacitor provides the AC current to the internal power MOSFETs.
- 3. Keep the switching node SW away from the sensitive V_{OUT} node.
- 4. Keep the (-) plates of C_{IN} and C_{OUT} as close as possible.

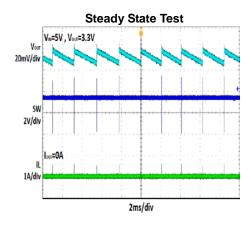


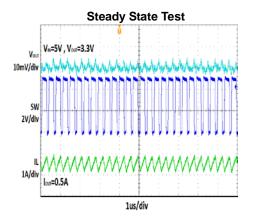


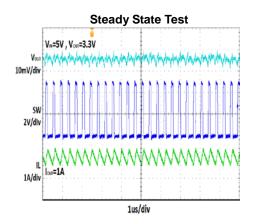


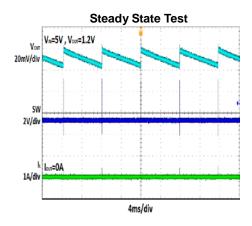
* TYPICAL CHARACTERISTICS

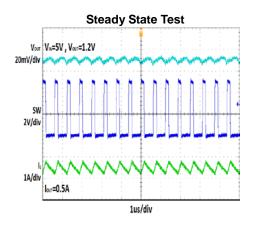
C_{IN}=10uF, C_{OUT}=10uF

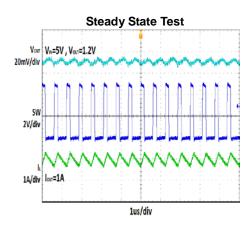






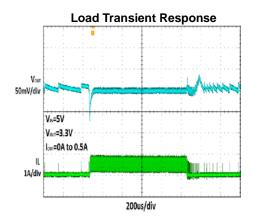


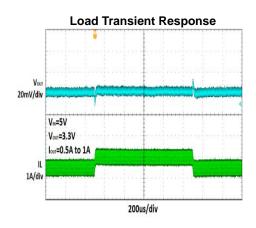


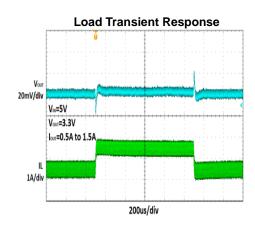


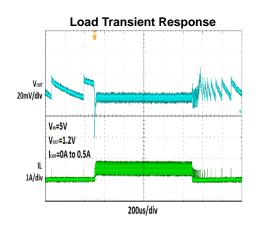


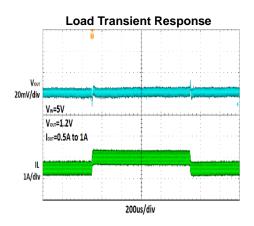
❖ TYPICAL CHARACTERISTICS (CONTINUOUS)

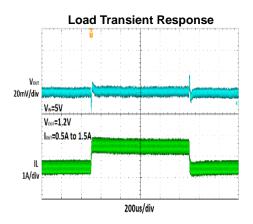






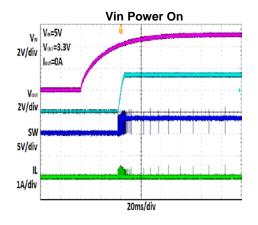


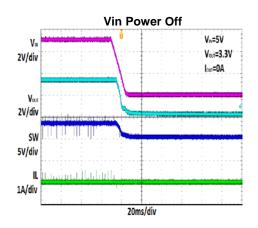


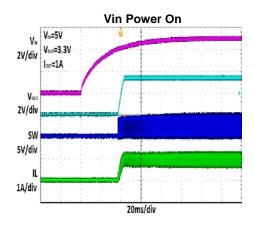


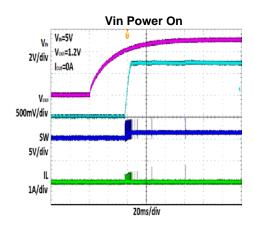


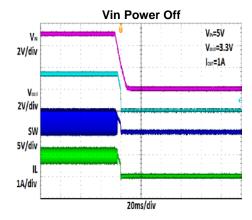
* TYPICAL CHARACTERISTICS (CONTINUOUS)

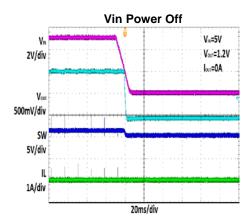




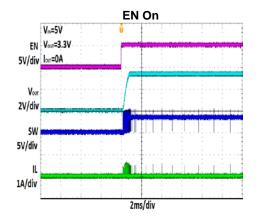


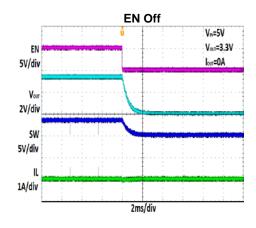


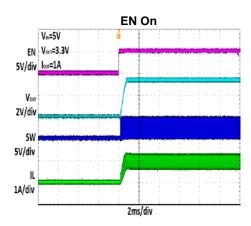


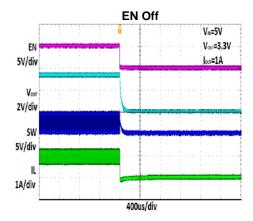


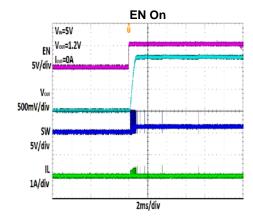
❖ TYPICAL CHARACTERISTICS (CONTINUOUS)

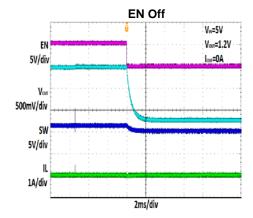




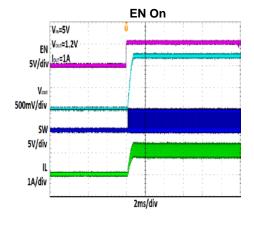


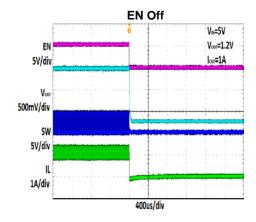


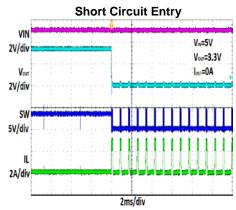


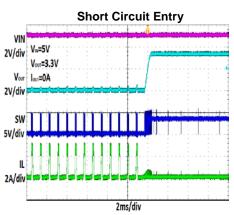


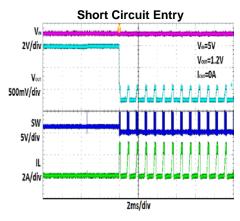
❖ TYPICAL CHARACTERISTICS (CONTINUOUS)

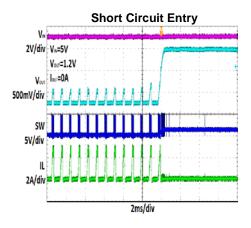


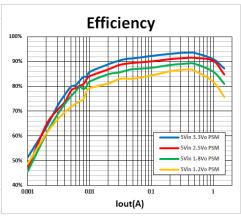










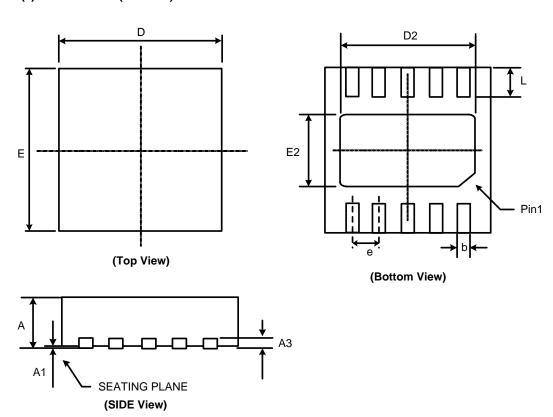


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❖ PACKAGE OUTLINES

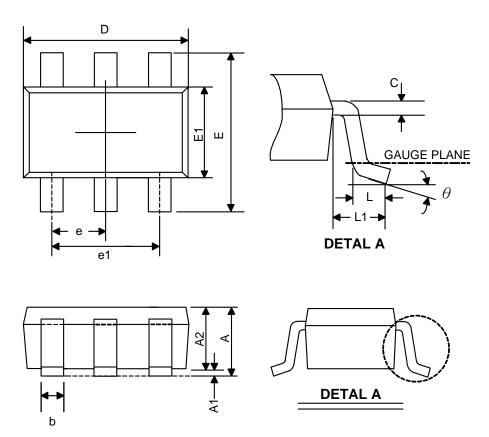
TDFN-10L (3*3mm²)



Symbol	Dimensions in Millimeters Dimensions in Inches						
Symbol	Min.	Nom.	Max.	Min.	Nom.	Max.	
А	0.7	0.75	0.8	0.028	0.03	0.031	
A1	0	0.02	0.05	0	0.001	0.002	
A3		0.20 REF.		0.008 REF.			
b	0.18	0.25	0.3	0.007	0.01	0.012	
D	2.9	3	3.1	0.114	0.118	0.122	
D2	2.2	2.4	2.5	0.087	0.094	0.098	
Е	2.9	3	3.1	0.114	0.118	0.122	
E2	1.5	1.6	1.7	0.059	0.063	0.07	
е		0.50 BSC.			0.020 BSC.		
L	0.3	0.4	0.5	0.012	0.016	0.02	



(2) SOT-23-6L

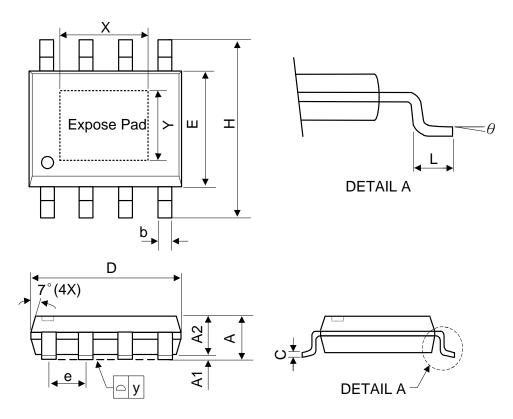


Cumbal	Dime	nsions in Millir	meters	Dimensions in Inches			
Symbol	Min.	Nom.	Max.	Min.	Nom.	Max.	
А	-	-	1.45	-	-	0.057	
A1	0	-	0.15	0	0.003	0.006	
A2	0.9	1.1	1.3	0.035	0.043	0.051	
b	0.3	0.4	0.5	0.012	0.016	0.02	
С	0.08	-	0.22	0.003	0.006	0.009	
D	2.7	2.9	3.1	0.106	0.114	0.122	
E1	1.4	1.6	1.8	0.055	0.063	0.071	
E	2.6	2.8	3	0.102	0.11	0.118	
L	0.3	0.45	0.6	0.012	0.018	0.024	
L1	0.5	0.6	0.7	0.02	0.024	0.028	
e1		1.9 BSC			0.075 BSC		
е		0.95 BSC			0.037 BSC		
θ	00	40	80	00	40	80	

JEDEC outline: MO-178 AB



(3) SOP-8L-EP



Cymabol	Dimensions in Millimeters			Dimensions in Inches		
Symbol	Min.	Nom.	Max.	Min.	Nom.	Max.
А	-	-	1.75	-	-	0.069
A1	0	-	0.15	0	-	0.06
A2	1.25	-	-	0.049	-	-
С	0.1	0.2	0.25	0.0075	0.008	0.01
D	4.7	4.9	5.1	0.185	0.193	0.2
E	3.7	3.9	4.1	0.146	0.154	0.161
Н	5.8	6	6.2	0.228	0.236	0.244
L	0.4	-	1.27	0.015	-	0.05
b	0.31	0.41	0.51	0.012	0.016	0.02
е		1.27 BSC		0	.050 BSC	
у	-	-	0.1	-	-	0.004
X	-	2.34	-	-	0.092	-
Υ	=	2.34	-	-	0.092	-
θ	00	-	80	00	-	80

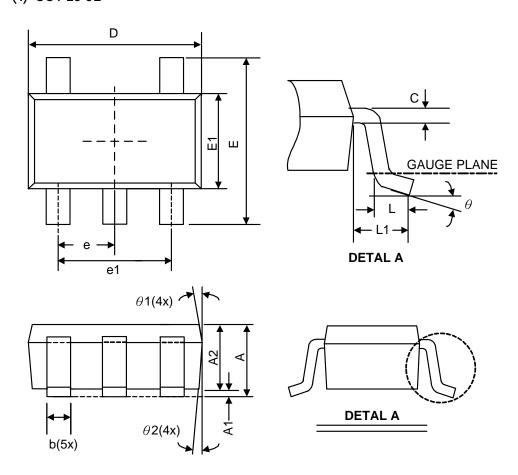
Mold flash shall not exceed 0.25mm per side

JEDEC outline: MS-012 BA

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(4) SOT-23-5L

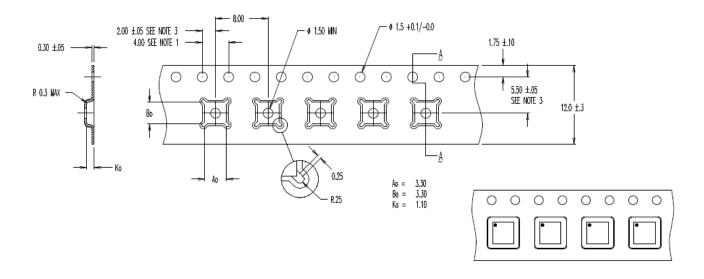


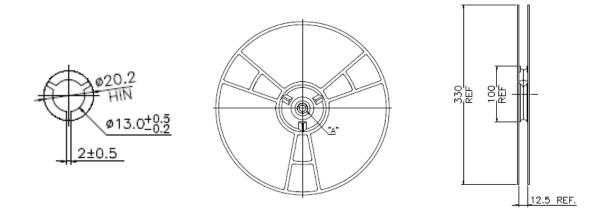
Symbol	Dime	nsions in Millir	neters	Dimensions in Inches			
Symbol	Min.	Nom.	Max.	Min.	Nom.	Max.	
А	-	-	1.45	-	-	0.057	
A1	0	0.08	0.15	0	0.003	0.006	
A2	0.9	1.1	1.3	0.035	0.043	0.051	
b	0.3	0.4	0.5	0.012	0.016	0.02	
С	0.08	0.15	0.22	0.003	0.006	0.009	
D	2.7	2.9	3.1	0.106	0.114	0.122	
E1	1.4	1.6	1.8	0.055	0.063	0.071	
E	2.6	2.8	3	0.102	0.11	0.118	
L	0.3	0.45	0.6	0.012	0.018	0.024	
L1	0.5	0.6	0.7	0.02	0.024	0.028	
e1		1.9 BSC			0.075 BSC		
е		0.95 BSC			0.037 BSC		
θ	00	40	80	00	40	80	
θ 1	5⁰	10°	15°	5°	10°	15º	
θ2	5°	10°	15º	5°	10º	15º	

JEDEC outline: MO-178 AA

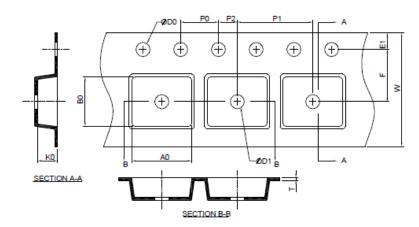
❖ TAPE & REEL CARRIER DIMENSIONS

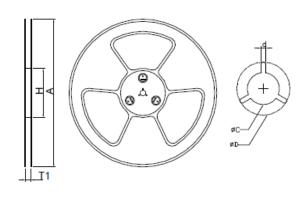
TDFN 3x3mm

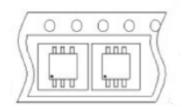




SOT-23-6L



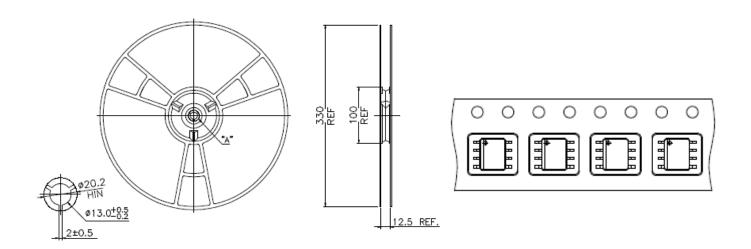


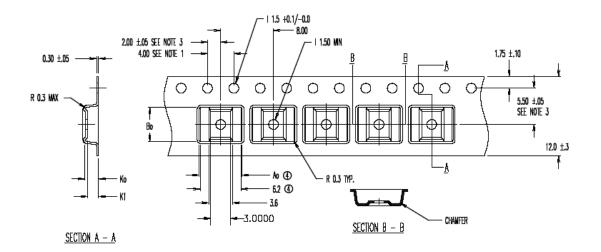


Α	Н	T1	С	d	D	W	E1	F
178.0±2.00	50 MIN.	8.4+2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	8.0±0.30	1.75±0.10	3.5±0.05
P0	P1	P2	D0	D1	T	A0	B0	K0
4.0±0.10	4.0±0.10	2.0±0.05	1.5+0.10	1.0 MIN.	0.6+0.00	3.20±0.20	3.10±0.20	1.50±0.20

(mm)

SOP-8L-EP



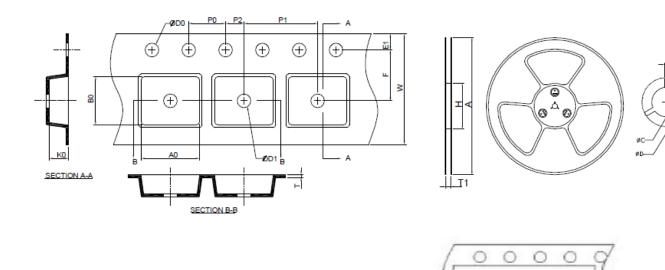


Notes:

- 1. 10 sprocket hole pitch cumulative tolerance ± 0.2mm
- 2. Camber not to exceed 1mm in 100mm.
- 3. Material: Anti-Static Black Advantek Polystyrene.
- 4. Ao and Bo measured on a plane 0.3mm above the bottom of the pocket.
 5. Ko measured from a plane on the inside bottom of the pocket to the top surface of the carrier.
- Pocket position relative to sprocket hole measured as true position of pocket, not pocket hole.



SOT-23-5L



Α	Н	T1	С	d	D	W	E1	F
178.0±2.00	50 MIN.	8.4+2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	8.0±0.30	1.75±0.10	3.5±0.05
D0								
P0	P1	P2	D0	D1	T	A0	B0	K0

(mm)